## RoHS Compliant \& Pb-Free Product

## Typical Applications

- CATV Distribution Amplifiers
- Cable Modems
- Broadband Gain Blocks
- Laser Diode Driver
- Return Channel Amplifier
- Base Stations


## Product Description

The RF2312 is a general purpose, low cost high linearity RF amplifier IC. The device is manufactured on an advanced Gallium Arsenide Heterojunction Bipolar Transistor (HBT) process, and has been designed for use as an easily cascadable $75 \Omega$ gain block. The gain flatness of better than 0.5 dB from 5 MHz to 1000 MHz , and the high linearity, make this part ideal for cable TV applications. Other applications include IF and RF amplification in wireless voice and data communication products operating in frequency bands up to 2500 MHz . The device is self-contained with $75 \Omega$ input and output impedances, and requires only two external DC biasing elements to operate as specified.

Optimum Technology Matching ${ }^{\circledR}$ Applied

| $\square$ Si BJT | $\square$ GaAs HBT | $\square$ GaAs MESFET |
| :--- | :--- | :--- |
| $\square$ Si Bi-CMOS | $\square$ SiGe HBT | $\square$ Si CMOS |
| $\square$ InGaP/HBT | $\square$ GaN HEMT | $\square$ SiGe Bi-CMOS |



Functional Block Diagram


## NOTES:

1. Shaded lead is pin 1.
2. All dimensions are excluding flash, protrusions or burrs.
3. Lead coplanarity: 0.005 with respect to datum " A ".
4. Package surface finish: Matte (Charmilles \#24~27).

Package Style: SOIC-8

## Features

- DC to well over 2500 MHz Operation
- Internally Matched Input and Output
- 15dB Small Signal Gain
- 3.8 dB Noise Figure
- +20 dBm Output Power
- Single 5V to 12V Positive Power Supply


## Ordering Information

| RF2312 | Linear General Purpose Amplifier |
| :--- | :--- |
| RF2312 PCBA | Fully Assembled Evaluation Board $-75 \Omega$ |
| RF2312 PCBA | Fully Assembled Evaluation Board $-50 \Omega$ |


| RF Micro Devices, Inc. | Tel (336) 664 1233 |
| :--- | ---: |
| 7628 Thorndike Road | Fax (336) 6640454 |
| Greensboro, NC 27409, USA | http://www.rfmd.com |

Greensboro, NC 27409, USA

## Absolute Maximum Ratings

| Parameter | Rating | Unit |
| :--- | :---: | :---: |
| Input RF Power | +18 | dBm |
| Output Load VSWR | $20: 1$ |  |
| Ambient Operating Temperature | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Storage Temperature | -40 to +150 | ${ }^{\circ} \mathrm{C}$ |

4 Caution! ESD sensitive device.
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RF2312

| Parameter | Specification |  |  | Unit | Condition |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. |  |  |
| Overall (75ת) | 14.5 | DC to 250016 | $\begin{aligned} & 4.3 \\ & 4.8 \end{aligned}$ | $\begin{gathered} \mathrm{MHz} \\ \mathrm{~dB} \\ \mathrm{~dB} \\ \mathrm{~dB} \end{gathered}$ | $\begin{aligned} & \mathrm{T}=25^{\circ} \mathrm{C}, \mathrm{~V}_{\mathrm{CC}}=9 \mathrm{~V}, \text { Freq }=900 \mathrm{MHz}, \\ & \mathrm{R}_{\mathrm{C}}=30 \Omega, 75 \Omega \text { System } \end{aligned}$ |
| Frequency Range |  |  |  |  | 3 dB Bandwidth |
| Gain |  |  |  |  |  |
| Noise Figure |  | 3.8 |  |  | From 50 MHz to $300 \mathrm{MHz},-30^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. |
| Input VSWR |  | 4.2 |  |  | From 300 MHz to $1000 \mathrm{MHz},-30^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. |
|  |  | 1.3:1 | 2:1 |  | From 50 MHz to $900 \mathrm{MHz},-30^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. Appropriate values for the DC blocking capacitors and bias inductor are required to maintain this VSWR at the intended operating frequency range. |
| Output VSWR |  | 1.2:1 | 1.75:1 |  | From 50 MHz to $300 \mathrm{MHz},-30^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. Appropriate values for the DC blocking capacitors and bias inductor are required to maintain this VSWR at the intended operating frequency range. |
|  |  | $1.4: 1$ $1.5: 1$ | $\begin{aligned} & \text { 2:1 } \\ & \text { 2:1 } \end{aligned}$ |  | From 300 MHz to $500 \mathrm{MHz},-30^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. From 500 MHz to $900 \mathrm{MHz},-30^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. |
| Output $\mathrm{IP}_{3}$ | +36 | +38 |  | dBm | At 100 MHz |
| Output $\mathrm{IP}_{3}$ | +33 | +36 |  | dBm | At 500 MHz |
| Output $\mathrm{IP}_{3}$ | +28 | +30 |  | dBm | At 900 MHz |
| Output $\mathrm{P}_{1 \mathrm{~dB}}$ | +21 | +22 |  | dBm | At 100 MHz |
| Output $\mathrm{P}_{1 \mathrm{~dB}}$ | +20 | +21 |  | dBm | At 500 MHz |
| Output $\mathrm{P}_{1 \mathrm{~dB}}$ | +17 | +18.5 |  | dBm | At 900 MHz |
| Saturated Output Power |  | +23 |  | dBm | At 100 MHz |
| Saturated Output Power |  | +22.5 |  | dBm | At 500 MHz |
| Saturated Output Power |  | +20.5 |  | dBm | At 900 MHz |
| Reverse Isolation |  | 20 |  | dB |  |
| 77 Channels |  |  |  |  | 77 Channels to 550 MHz at 10 dBmV , 33 channels to 760 MHz at 0 dBmV flat at DUT input |
| cso |  | >86 |  | dBc | 61.25 MHz |
|  |  | >86 |  | dBc | 83.25 MHz |
|  |  | 76 |  | dBc | 193.25 MHz |
|  |  | 72 |  | dBc | 313.2625 MHz |
|  |  | 64 |  | dBc | 547.25 MHz |
| CTB |  | >86 |  | dBc | 61.25 MHz |
|  |  | >86 |  | dBc | 83.25 MHz |
|  |  | 86 |  | dBc | 193.25 MHz |
|  |  | 84 |  | dBc | 313.2625 MHz |
|  |  | 83 |  | dBc | 547.25 MHz |
| CNR 110 Channels | 65 | 66 |  | dB |  |
|  | 65 |  |  |  | 110 Channels, 10dBmV/channel at input |
| CSO |  | >86 |  | dBc | 61.25 MHz |
|  |  | >86 |  | dBc | 83.25 MHz |
|  |  | 76 |  | dBc | 193.25 MHz |
|  |  | 70 |  | dBc | 313.2625 MHz |
|  |  | 64 |  | dBc | 547.25 MHz |
| СTB |  | 84 |  | dBc | 61.25 MHz |
|  |  | 86 |  | dBc | 83.25 MHz |
|  |  | 85 |  | dBc | 193.25 MHz |
|  |  | 81 |  | dBc | 313.2625 MHz |
|  |  | 80 |  | dBc | 547.25 MHz |
| Cross Modulation |  | 77 |  | dBc | 61.25 MHz |
|  |  | 74 |  | dBc | 445.25 MHz |
| CNR |  | 66 |  | dB |  |


| Parameter | Specification |  |  | Unit | Condition |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | Min. | Typ. | Max. |  |  |
| Overall (75ת Push-Pull) |  |  |  |  | $\begin{aligned} & \mathrm{T}=25^{\circ} \mathrm{C}, \mathrm{~V}_{\mathrm{CC}}=9 \mathrm{~V} \text { or } 24 \mathrm{~V}, 75 \Omega \text { System, } \\ & \mathrm{RF}_{\mathrm{IN}}=-10 \mathrm{dBm} \end{aligned}$ |
| Frequency Range |  | DC to 150 |  | MHz |  |
| Gain |  | 15 |  | dB |  |
| Noise Figure |  | 5.0 |  | dB | From 5 MHz to $150 \mathrm{MHz},-30^{\circ} \mathrm{C}$ to $+70^{\circ} \mathrm{C}$. |
| Input VSWR |  | 1.1:1 |  |  |  |
| Output VSWR |  | 1.2:1 |  |  |  |
| Output $\mathrm{IP}_{2}$ |  | +71 |  | dBm | At 10 MHz |
|  |  | +72 |  | dBm | At 30 MHz |
|  |  | +74 |  | dBm | At 50 MHz |
| Output $\mathrm{IP}_{3}$ |  | +40 |  | dBm | At 10 MHz |
|  |  | +40 |  | dBm | At 30 MHz |
|  |  | +40 |  | dBm | At 50 MHz |
| Second Harmonic |  | -73 |  | dBc | At 10 MHz |
|  |  | -65 |  | dBc | At 30 MHz |
|  |  | -65 |  | dBc | At 50 MHz |


| Pin | Function | Description | Interface Schematic |
| :---: | :---: | :---: | :---: |
| 1 | RF IN | RF input pin. This pin is NOT internally DC-blocked. A DC-blocking capacitor, suitable for the frequency of operation, should be used in all applications. The device has internal feedback, and not using a DCblocking capacitor will disable the temperature compensation. The bias of the device can be controlled by this pin. Adding an optional $1 \mathrm{k} \Omega$ resistor to ground on this pin reduces the bias level, which may be compensated for by a higher supply voltage to maintain the appropriate bias level. The net effect of this is an increased output power capability, as well as higher linearity for signals with high crest factors. DC-coupling of the input is not allowed, because this will override the internal feedback loop and cause temperature instability. |  |
| 2 | GND | Ground connection. For best performance, keep traces physically short and connect immediately to ground plane. Each ground pin should have a via to the ground plane. |  |
| 3 | GND | Same as pin 2. |  |
| 4 | GND | Same as pin 2. |  |
| 5 | GND | Same as pin 2. |  |
| 6 | GND | Same as pin 2. |  |
| 7 | GND | Same as pin 2. |  |
| 8 | RF OUT | RF output and bias pin. Because DC is present on this pin, a DC-blocking capacitor, suitable for the frequency of operation, should be used in most applications. For biasing, an RF choke in series with a resistor is needed. The value for the resistor $R_{C}$ is $30 \Omega(0.5 \mathrm{~W})$ for $\mathrm{V}_{C C}=9 \mathrm{~V}$ and $21 \Omega$ for $\mathrm{V}_{\mathrm{CC}}=8 \mathrm{~V}$. The DC voltage on this pin is typically 6.0 V with a current of 100 mA . In lower power applications the value of $R_{C}$ can be increased to lower the current and $V_{D}$ on this pin. |  |

## Application Schematic 5 MHz to 50 MHz Reverse Path



NOTE 1:
Optional resistor $R_{S}$ can be used to maintain the correct bias level at higher supply voltages. This is used to increase output capability or linearity for signals with high crest factors.

## Application Schematic 10dB Gain



R5 is used to maintain the correct bias level at higher supply voltages and is also required in this configuration. The RC network of R2 and C3 should be kept physically as short as possible. R2 can be adjusted as required to improve the impedance matching. R6 and R7 reduce the typical gain by increasing the emitter resistance. L1 should be at least $200 \Omega$ reactive at the lowest operating frequency. C1 and C 2 should be less than $10 \Omega$ at the lowest operating frequency. C4 and C 5 improve gain flatness.

## Application Schematic Push-Pull Standard Voltage



## Application Schematic Push-Pull 24 V



## Evaluation Board Schematic - $50 \Omega$

(Download Bill of Materials from www.rfmd.com.)


Evaluation Board Schematic - $75 \Omega$


NOTE: For 5 V applications, R1 to R4 may be removed (shorted). This will result in degraded distortion performance.

## Evaluation Board Layout - $50 \Omega$ <br> 2.02" x 2.02"

Board Thickness 0.031", Board Material FR-4


## Evaluation Board Layout - 75 $\Omega$ <br> Standard Voltage <br> $1.40 " \times 1.40 "$ <br> Board Thickness 0.062", Board Material FR-4



## Evaluation Board Layout - $75 \Omega$ <br> Push-Pull, Standard Voltage <br> $1.70 "$ x $1.50 "$ <br> Board Thickness 0.062", Board Material FR-4



Evaluation Board Layout - 75<br>Push-Pull, 24V<br>$1.70^{\prime \prime} \times 1.50 "$<br>Board Thickness 0.062", Board Material FR-4




Output Third Order Intercept Point (OIP3) versus $P_{\text {IN }}$


IM3 Products versus $P_{\text {out }}$



Output P1dB versus Frequency








75 Ohms, ICC $=100 \mathrm{~mA}, \mathrm{Temp}=\mathbf{2 5}{ }^{\circ} \mathrm{C}$


75 Ohms, $\mathrm{ICC}=\mathbf{1 1 0} \mathrm{mA}, \mathrm{Temp}=\mathbf{2 5}{ }^{\circ} \mathrm{C}$


